B.TECH/EE/6TH SEM/ ELEC 3231/2019 DIGITAL SIGNAL PROCESSING (ELEC 3231)

Time Allotted : 3 hrs

Full Marks : 70

Figures out of the right margin indicate full marks.

Candidates are required to answer Group A and <u>any 5 (five)</u> from Group B to E, taking <u>at least one</u> from each group.

Candidates are required to give answer in their own words as far as practicable.

Group – A (Multiple Choice Type Questions)

- 1. Choose the correct alternative for the following: $10 \times 1 = 10$
 - (i) The common period (N) of the signal $x(n) = \cos(0.4\pi n) + \sin(0.5\pi n + 30^{\circ})$ is (a) N = 20 (b) N = 5 (c) N = 4 (d) N = 2.
 - (ii) From a given signal x(n), a signal x(-n α) can be generated using one of the following sequence of operations.
 (a) time reversal of x(n) and then delayed by α
 (b) x(n) delayed by α and then time reversal
 - (c) x(n) advanced by α and then time reversal
 - (d) time reversal of x(n) and then upscaling by α .
 - (iii) The value of y(2) using linear convolution (y(n) = x(n) * h(n)) of the sequences $x(n) = (3, 2, 4), 0 \le n \le 2$ and $h(n) = (-2, 3), 0 \le n \le 1$ is (a)12 (b) 5 (c) -2 (d) -6.
 - (iv) The z -transform of a 'left sided' signal $X(z) = \frac{z}{z-b}$ convergences if ROC is (a) inside an annulus bounded by $|b| < |z| < \infty$ (b) outside a circle of radius |z| > |b|(c) on a circle of radius |z| = |b|(d) inside a circle of radius |z| < |b|.
 - (v) If the z transform of a signal (x(n)) is $X(z) = \frac{z}{z-0.5}$; ROC: |z| > 0.5, then z - transform of $\overline{x}(n) = nx(n)$ is (a) $\overline{X}(z) = \frac{1}{z(z-0.5)}$; ROC: |z| < 0.5(b) $\overline{X}(z) = \frac{(z-1)}{(z-0.5)}$; ROC: 0.5 < |z| < 1(c) $\overline{X}(z) = \frac{(z-0.5)}{1}$; ROC: |z| > 0.5(d) $\overline{X}(z) = \frac{0.5z}{(z-0.5)^2}$; ROC: |z| > 0.5.

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(vii)

(vi) A periodic signal x(n) is expressed as $x(n) = 1 + 0.666 \cos(\frac{2\pi}{3}n + 120^{0})$, the corresponding normalized power signal P_x is (a)1.666 (b) 1.444 (c) 1.105 (d) 1.222.

A periodic signal x(n) (with fundamental period N = 3) is expressed in exponential form $x(n) = \sum_{k=0}^{2} D_k e^{j(\frac{2\pi}{3})kn}$ where, $D_0 = 1$, $D_1 = 0.333 \angle 120^\circ$, and $D_2 = 0.333 \angle -120^\circ$. The time domain signal x(n) can be expressed from DTFT coefficients as

(a)
$$x(n) = 1 + 0.333 \sin\left(\frac{\pi n}{3} + 120\right)$$

(b) $x(n) = 0.707 + 0.235 \cos\left(\frac{2\pi n}{3} - 120\right)$
(c) $x(n) = 1 + 0.666 \cos\left(\frac{2\pi}{3}n + 120\right)$
(d) $x(n) = 1.33 + 0.666 \sin\left(\frac{2\pi n}{3} - 120\right)$.

- (viii) The DFT coefficient X(2) of the four point segment x(0) = 0, x(1) = 1, x(2) = 2, x(3) = 3 of a sequence x(n) is (a) X(2) = 6 (b) X(2) = -2 (c) X(2) = -2 + j2 (d) X(2) = -2 - j2.
- (ix) The equivalent digital filter H(z) for the given analog filter H_c(s) = $\frac{1}{s+1}$ using impulse invariance mapping with sampling time T_s = 1s is (a) H(z) = $\frac{z}{z-0.368}$ (b) H(z) = $\frac{z-1}{z-0.368}$ (c) H(z) = $\frac{z}{z-0.638}$ (d) H(z) = $\frac{z}{z-1}$.
- (x) If a causal and stable discrete time system $H(z) = \frac{z}{z-0.8}$ is excited with a sinusoidal input $x(n) = \cos(0.05\pi n) u(n)$ then the amplitude (B) of the sinusoidal output response $y_{ss}(n) = B\cos(0.05\pi n + \theta)$ at steady-state is (a) 6 (b) 4.1 (c) 4.5 (d) 2.5

Group – B

- 2. (a) A non-recursive filter of length (M + 1) is described by $y(n) = b_0 x(n) + b_1 x(n-1) + b_2 x(n-2) + \dots + b_M x(n-M)$. Show that the impulse response sequence or samples are the filter 'tap-coefficients' (i.e. $b_i = 0, 1, 2, \dots M$).
 - (b) (i) If x(n) is the input and h(n) is the impulse response of a linear timeinvariant discrete system 'S', then the show that the output y(n) is given by $y(n) = \sum_{m=0}^{\infty} x(m)h(n-m)$ (assume both the signals x(n), & h(n) are causal).
 - (ii) Convert a non-recursive filter y(n) = x(n) + x(n-1) + x(n-2) to an equivalent recursive form and hence compute recursively the values of y(n) for n = 0 to 5 with $x(n) = \delta(n)$ and y(-1) = 0. Comment on the length of the output sequence.

3 + (5 + 4) = 12

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- 3. (a) Prove that linear time invariant system with impulse response h(n) is stable, in the bounded-input bounded-output sense, if and only if the impulse response is absolutely summable, that is, if $\sum_{n=-\infty}^{\infty} |h(n)| < \infty$.
- (b) Realize the following digital filter using "direct-form-II" structure. $y(n) = \frac{5}{4}y(n-1) - \frac{3}{4}y(n-2) + \frac{1}{8}y(n-3) + 8x(n) - 4x(n-1) + 11x(n-2) - 2x(n-3).$ Is the digital filter stable?

Group – C

- 4. (a) Let X(z) = Z(x(n)) with ROC: R_x , if the signal x(n), $n \ge 0$ is multiplied by n, then show that $Z(nx(n)) = -z \frac{dX(z)}{dZ}$, with ROC: R_x .
- (b) Find the inverse Z-transform of $H(z) = \frac{1}{(1-z^{-1})(1-0.5z^{-1})}$, and its ROC. Assume h(n) is left-sided signal.

5 + 7 = 12

5 + 7 = 12

9. (a)

- 5. (a) (i) What is the basic principle behind the impulse-invariance transformation?
 - (ii) How does the s plane get mapped into the z plane under the impulse-invariance transformation?—Discuss.
 - (b) Transform a filter $H(s) = \frac{s+1}{s^2+5s+6}$, into an equivalent digital filter H(z) using the impulse invariance technique in which $T_s = 0.1s$ and hence express H(z) as an equivalent difference equation for computer simulation.

(2+3)+7=12

Group - D

- 6. (a) Define the DFT of a finite sequence. Show that the N point DFT of a finite sequence signal x(n) is periodic with period 'N'.
- (b) Given a sequence $x(n) = \{1, 2, 1, 0\}$ for $0 \le n \le 3$, evaluate DFT of a shifted signal y(n) = x(n 1), i.e. Y(k). Assume that $f_s = 100$ Hz. Sketch the amplitude spectrum, phase spectrum and power spectrum in context with the signal y(n).

4 + 8 = 12

- 7. (a) (i) What is circular convolution, why should we care about it, and how is it different from the linear convolution?
 - (ii) If $x(n) = \{6, 2, a, 0, b\}$, for n = 0, 1, 2, 3, 4 is circularly even symmetric, find the values of 'a' and 'b'.
 - (b) Let $x(n) = \{1, -1, -1, 1\}$ for n = 0, 1, 2, 3, and $h(n) = \{1, 2, 2, 1\}$, for n = 0, 1, 2, 3. Determine the output response y(n) using-circular convolution $(x(n) \otimes h(n))$.

Group – E

- 8. (a) (i) Show that any rational system H(z) can be decomposed into a minimum-phase system and an all-pass system.
 - (ii) It is said that IIR filters cannot have linear phase. Do you agree or disagree? Explain.
 - (b) The impulse response of a FIR filter is given as $h(n) = \{2, 1, 0, -1, -2\}$, for n = 0, 1, 2, 3, 4. Is it a linear-phase filter? If so, what type? Obtain the transfer function of the FIR filter (system) and sketch the pole-zero distributions of the FIR filter. (2 +3) + 7 = 12

A lowpass FIR filter having the following frequency specifications: Passband edge frequency $\Omega_p = 0.375\pi$ rad/s; stopband edge frequency, $\Omega_s = 0.5\pi$ rad/s; cut-off frequency, $\Omega_c = 0.438\pi$ rad/s; transition band, (width)=0.125\pi rad/s; passband ripple, $\delta_p \leq 0.0575$, stopband, $\delta_s \leq 0.0032$. Sketch the tolerance diagram for the low pass-filter

(b) Design a low-pass FIR filter with the specifications given in Q.9 (a) using the "Hamming Window" technique.

indicating all design specifications.

4 + 8 = 12

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